

# **Computer-Aided Extraction of Small-Signal Model Parameters of Heterojunction Bipolar Transistors**

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- ❑ An approach to **parameter extraction of small-signal model of heterojunction bipolar transistors (HBT's)** is proposed in the present paper based on *S*-parameter measurements.
- ❑ **Exact closed-form equations** are used for the direct extraction of circuit elements. The method is characterized by its simplicity and ease of implementation.
- ❑ The *Cadence PSpice* simulator and the graphical analyzer *Cadence Probe* are used to automate the extraction procedure.
- ❑ The parameter extraction of intrinsic elements of the model is realized using **macro-definitions in Probe**. The corresponding macros in the graphical analyzer Probe are presented realizing the proposed extraction procedure.



# Description of the Parameter Extraction Procedure

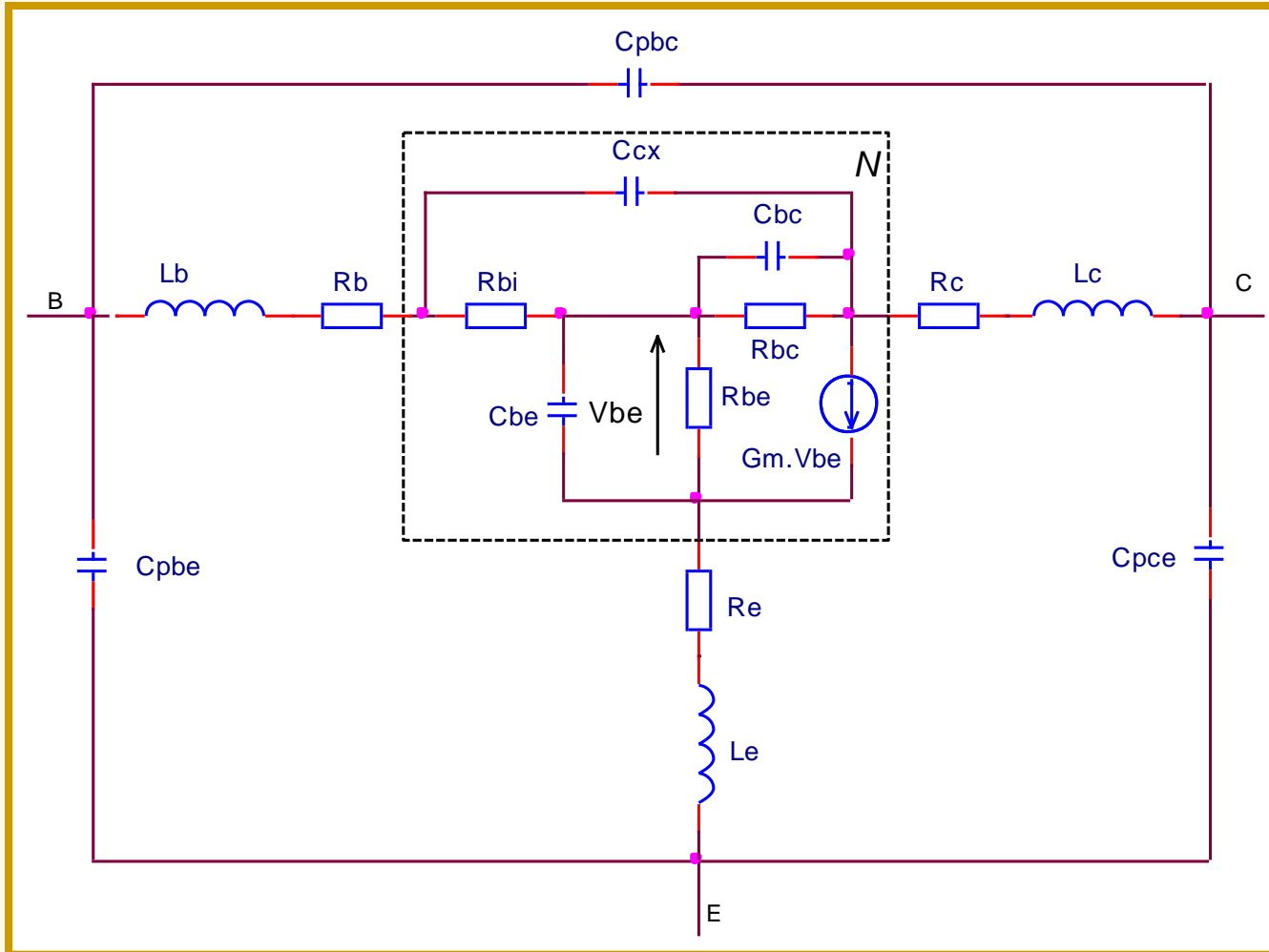


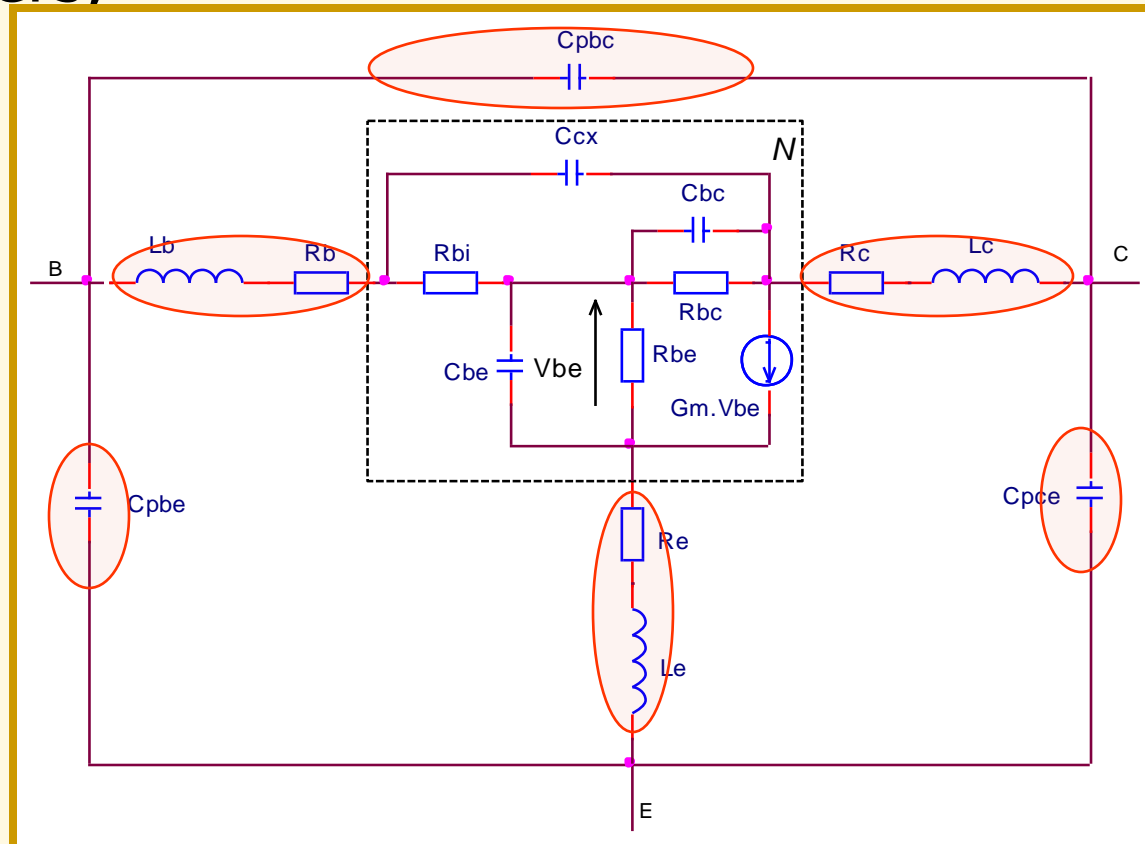
Fig. 1. Small-signal model of a HBT



# Description of the Parameter Extraction Procedure

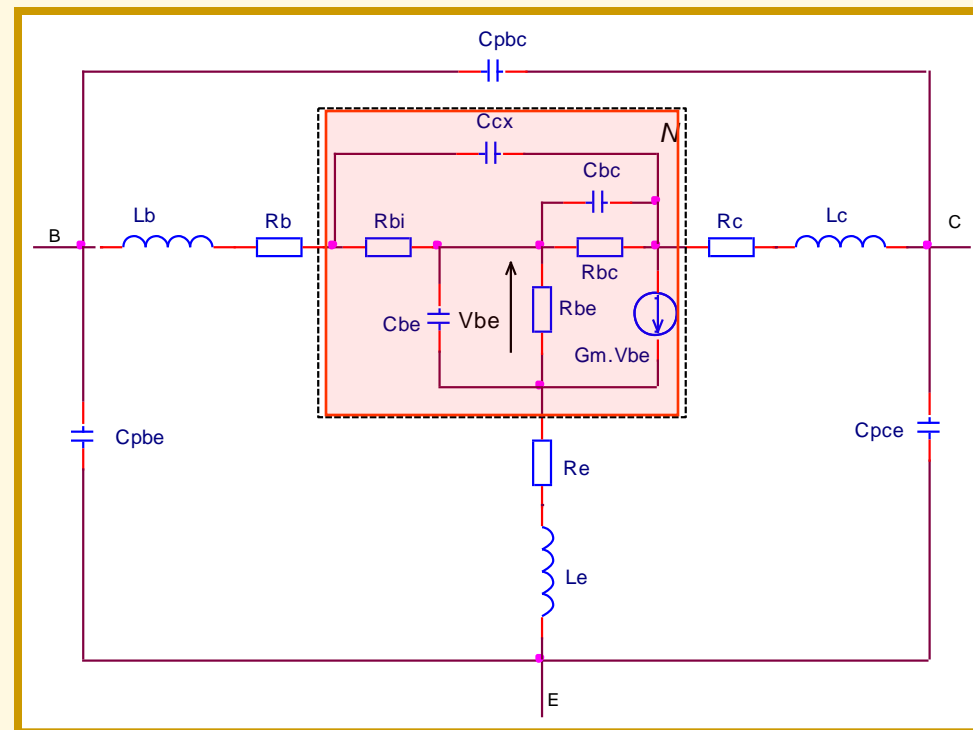
The direct extraction procedure consists of the following steps:

- De-embedding parasitic capacitances, inductances and resistances from the measured two-port  $S$ -parameters;



# Description of the Parameter Extraction Procedure

- ❑ Conversion of the measured  $S$  to  $Y$ -two-port parameters;
- ❑ Determination of  $R_{be}$  and  $C_{be}$ ;
- ❑ Determination of  $R_{bi}$ ;
- ❑ Determination of  $R_{bc}$  and  $C_{bc}$ ;
- ❑ Determination of  $C_{cx}$ ,  $G_{mo}$  and  $\tau$ .



## A. Introducing the $S$ -parameters

- The  $S$ -parameters are defined using voltage controlled current sources **VCCS** of GFREQ type (Fig. 2). The  $S$ -parameters are defined using voltage controlled current sources VCCS of GFREQ type.

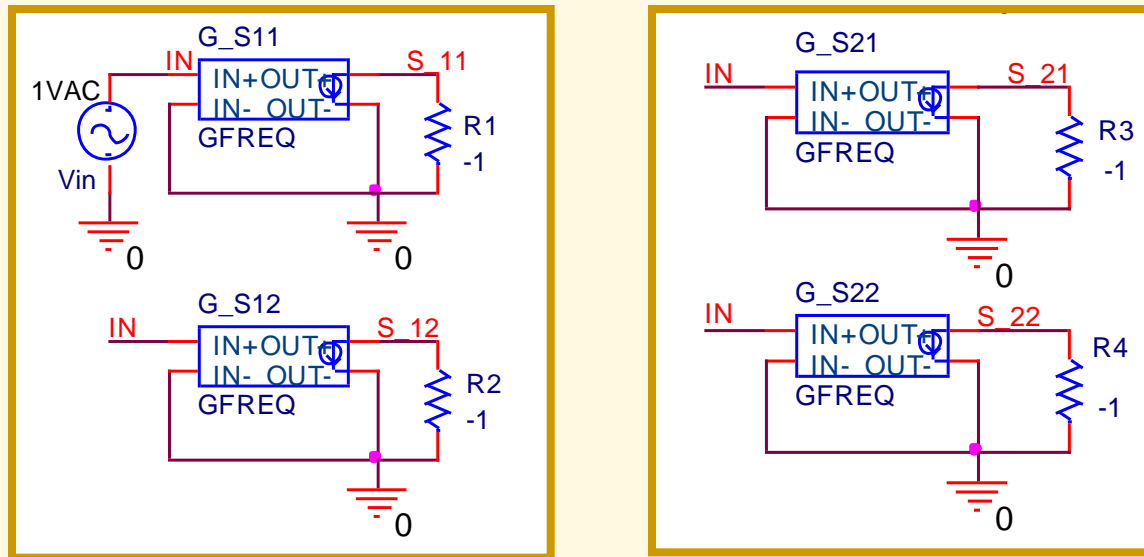


Fig. 2



## ***A. Introducing the S-parameters***

- ❑ The table has the form:

<frequency,magnitude [DB], phase [DEG]>:  
(freq1, mag1, phase1), (freq2, mag2, phase2) ...

- ❑ AC analysis is performed in the defined frequency range. The S-parameters are obtained using the following macros in the graphical analyzer Probe:

$$S_{11} = V(S_{11})$$

$$S_{21} = V(S_{21})$$

$$S_{12} = V(S_{12})$$

$$S_{22} = V(S_{22})$$



## ***B. Conversion of S- to Y- two-port parameters***

$$Y_{11} = \frac{(1 - S_{11})(1 + S_{22}) + S_{12}S_{21}}{\Delta} \quad (1)$$

$$Y_{22} = \frac{(1 + S_{11})(1 - S_{22}) + S_{12}S_{21}}{\Delta} \quad (2)$$

$$Y_{12} = \frac{-2S_{12}}{\Delta} \quad Y_{21} = \frac{-2S_{21}}{\Delta} \quad (3)$$

$$\Delta = ((1 + S_{11})(1 + S_{22}) - S_{12}S_{21})R_o$$





## ***B. Conversion of S- to Y- two-port parameters***

The conversion is performed using the following macros in the graphical analyzer *Probe*:

$$R_o=50$$

$$\text{delta} = ((1+S_{11})*(1+S_{22})-S_{12}*S_{21})*R_o$$

$$Y_{11} = ((1-S_{11})*(1+S_{22})+S_{12}*S_{21})/\text{delta}$$

$$Y_{12} = -2*S_{12}/\text{delta}$$

$$Y_{21} = -2*S_{21}/\text{delta}$$

$$Y_{22} = ((1+S_{11})*(1-S_{22})+S_{12}*S_{21})/\text{delta}$$



The intrinsic part N of the small-signal transistor model of a HBT is shown in Fig. 3. Applying T- $\Pi$  transformation for the subcircuit N1, the equivalent circuit shown in Fig. 4 is obtained.

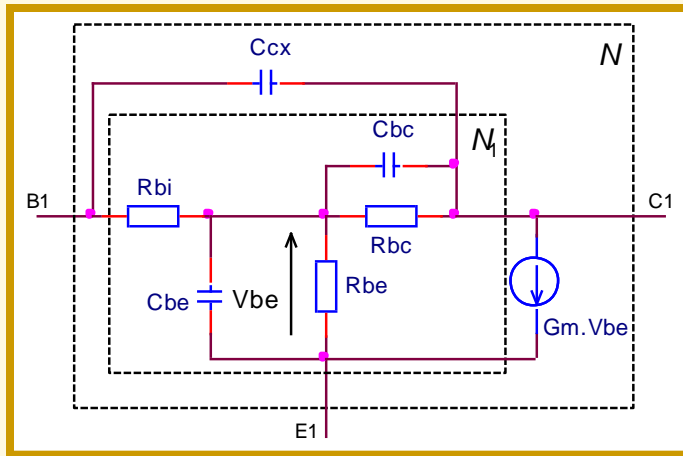


Fig. 3

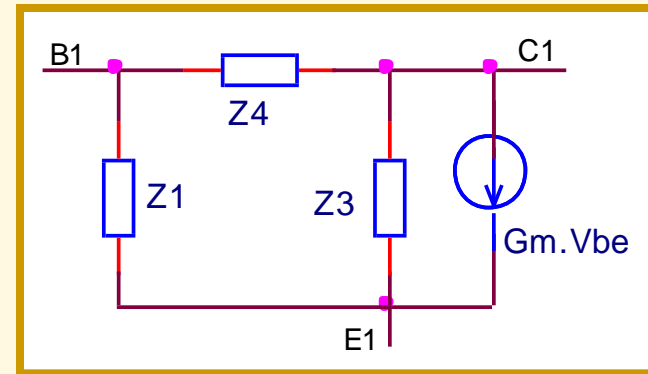


Fig. 4

$$Z_1 = \frac{D}{Z_{bc}} ; Z_2 = \frac{D}{Z_{be}} ; Z_3 = \frac{D}{Z_{bi}} , \quad (4)$$

$$Z_{bi} = R_{bi} ; Z_{bc} = \frac{R_{bc}}{1 + j\omega R_{bc} C_{bc}} ; Z_{be} = \frac{R_{be}}{1 + j\omega R_{be} C_{be}} , \quad (5)$$

$$D = Z_{bi} Z_{bc} + Z_{be} Z_{bc} + Z_{bi} Z_{be} \quad (6)$$

$$Z_4 = \frac{Z_2 Z_{cx}}{Z_2 + Z_{cx}} ; Z_{cx} = \frac{1}{j\omega C_{cx}} \quad (7)$$

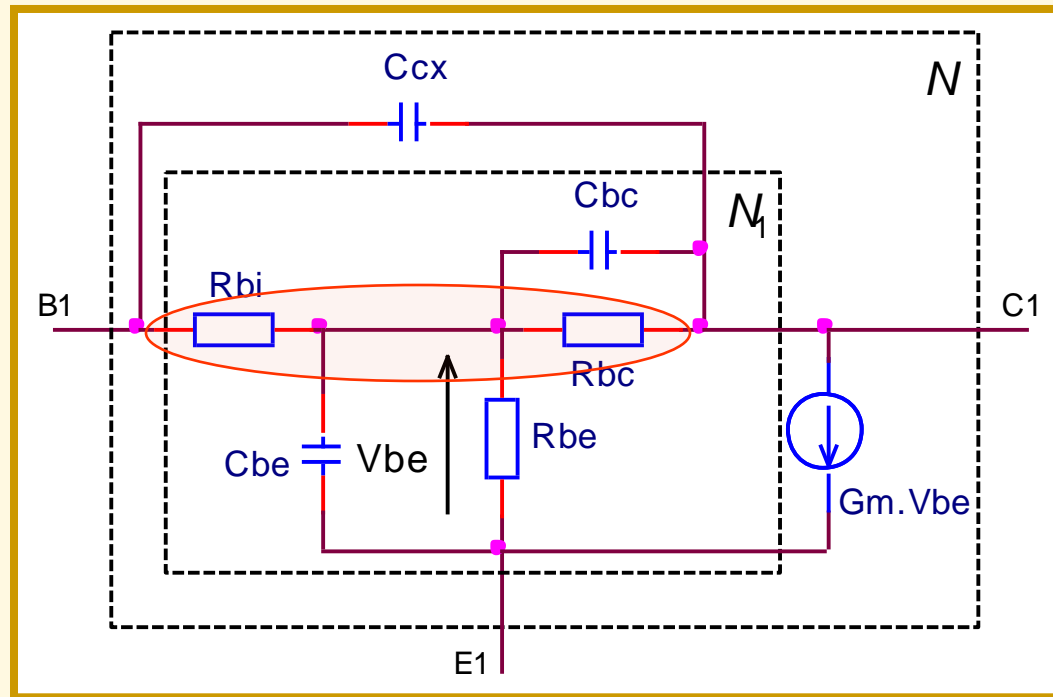
- The parameters  $Z_1$ ,  $Z_2$  and  $Z_3$  are obtained from the  $Y$ - parameters:

$$Z_1 = \frac{1}{Y_{11} + Y_{12}} ; \quad Z_4 = -\frac{1}{Y_{12}},$$

$$Z_3 = \frac{Y_{21} + Y_{11}}{(Y_{11} + Y_{12}) \cdot (Y_{22} + Y_{12})}.$$



## C. Determination of $(R_{bi}/R_{bc})$ and $(R_{bi}C_{bc})$



$$K_{Z13} = \frac{Z_1}{Z_3} = \frac{R_{bi}}{R_{bc}} (1 + j\omega R_{bc} C_{bc})$$

$$\frac{R_{bi}}{R_{bc}} = \operatorname{Re} \left( \frac{Z_1}{Z_3} \right)$$

$$K_{Z13} = \frac{Z_1}{Z_3} = \frac{Y_{22} + Y_{12}}{Y_{21} + Y_{11}}$$

$$R_{bi} C_{bc} = \frac{1}{\omega} \operatorname{Im} \left( \frac{Z_1}{Z_3} \right)$$

(11)

- ❑ The extraction procedure is realized using macro-definitions in the graphical analyzer *Cadence Probe*.
- ❑ The correspondence of the model parameter names and the names used in *Probe*, is given in Table I.

Table I

parameter name	parameter name in Probe	parameter name	parameter name in Probe
$R_{be}$	Rbe	$\tau$	TAU
$R_{bc}$	Rbc	$C_{cx}$	Ccx
$C_{be}$	Cbe	$R_{bIBC} = R_{bi} / R_{bc}$	RbIBC
$C_{bc}$	Cbc	$R_{bi} C_{bc}$	RbiCbc
$G_{mo}$	Gm0	$K_{Z13} = Z_1 / Z_3$	KZ13
$G_m$	Gm	$T_{be} = R_{be} C_{be}$	Tbe
$R_{bi}$	Rbi	$R$	RR

Determination of the ratio  $(R_{bi}/R_{bc})$  using the *Probe* macros:

$$KZ13=(Y22+Y12)/(Y21+Y11)$$

$$R_{bIBC\_f} = R(Z13)$$

$$R_{bIBC} = \max(\text{rms}(R_{bIBC\_f}))$$

Determination of the product  $(R_{bi} \cdot C_{bc})$  using the macros:

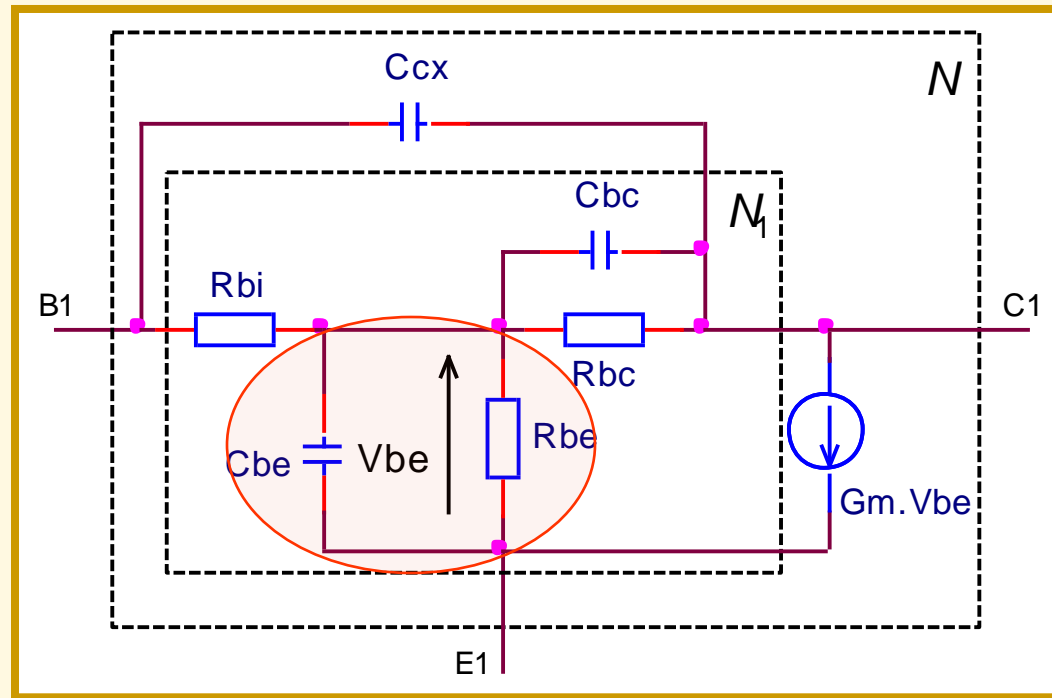
$$w=2*\pi*\text{Frequency}$$

$$R_{biCbc\_f} = \text{img}(Z13)/w$$

$$R_{biCbc} = \max(\text{rms}(R_{biCbc\_f}))$$



## D. Determination of $(R_{be}, C_{be})$



Using (4) and (5), the following expression for  $Z_1$  is obtained:

$$Z_1 = \frac{R(1 + j\omega T)}{1 + j\omega T_{be}}$$

$$R = R_{bi} R_{be} \left[ \frac{1}{R_{bc}} + \frac{1}{R_{be}} + \frac{1}{R_{bi}} \right]$$

(13)

$\text{Im}(Z_1)$  is obtained from (13):

$$\text{Im}(Z_1) = \frac{R(T - T_{be})\omega}{1 + \omega^2 T_{be}^2};$$

$$F_1 = \frac{\omega}{\text{Im}(Z_1)} = A + \omega^2 B;$$

$$A = \frac{1}{R(T - T_{be})}; \quad B = \frac{T_{be}^2}{R(T - T_{be})}.$$

$A$  and  $B$  are determined using least squares fitting of  $F_1$  as a function of the frequency.

$T_{be}$  is obtained in the form:

$$R_{be} C_{be} = T_{be} = \sqrt{\frac{B}{A}}.$$

(19)



Determination of the product (Rbe.Cbe) in *Probe*:

$$Z1 = 1 / (Y11 + Y12)$$

$$F1 = w / \text{IMG}(Z1)$$

$$F_{\text{max}} = \text{max}(\text{frequency})$$

$$w_{\text{max}} = 2 * \text{pi} * F_{\text{max}}$$

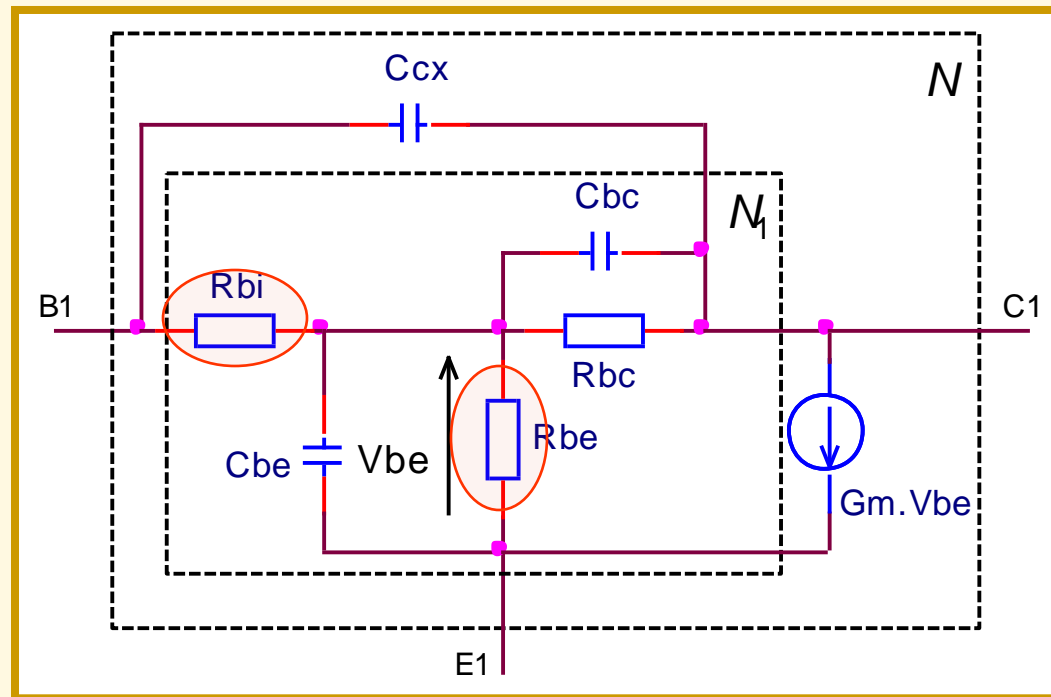
$$B = \text{MIN}(F1) / w_{\text{max}} / w_{\text{max}}$$

$$A = \text{min}(F1 - w * w * B)$$

$$T_{\text{be}} = \text{SQRT}(B / A)$$



## *E. Determination of $R_{be}$ and $R_{bi}$*



$$F_2 = Z_1 (1 + j\omega T_{be}) = R(1 + j\omega T)$$

$$R = \text{Re}(F_2)$$

$$T = \frac{\text{Im}(F_2)}{\omega R}$$

$R$  and  $T$  are determined applying least squares fitting of and as a function of the frequency using the following macros in *Probe* :

```

j=sqrt(-1)
F2=Z1*(1+j*w*Tbe)
RR_f=R(F2)
RR = max(rms(RR_f))
T_f=IMG(F2)/w/RR
T=max(rms(T_f))
    
```

Solving the linear system:

$$\begin{cases} R = CR_{be} + R_{bi} \\ RT = R_{bi}C_{bc}R_{be} + T_{be}R_{bi} \end{cases}$$

$$C = \begin{pmatrix} 1 + \frac{R_{bi}}{R_{bc}} \\ \frac{R_{bi}}{R_{bc}} \end{pmatrix}$$

$R_{be}$  and  $R_{bi}$  are determined:

$$R_{be} = \frac{R(T - T_{be})}{R_{bi}C_{bc} - T_{be}C}$$

$$R_{bi} = R - CR_{be} \quad (23)$$

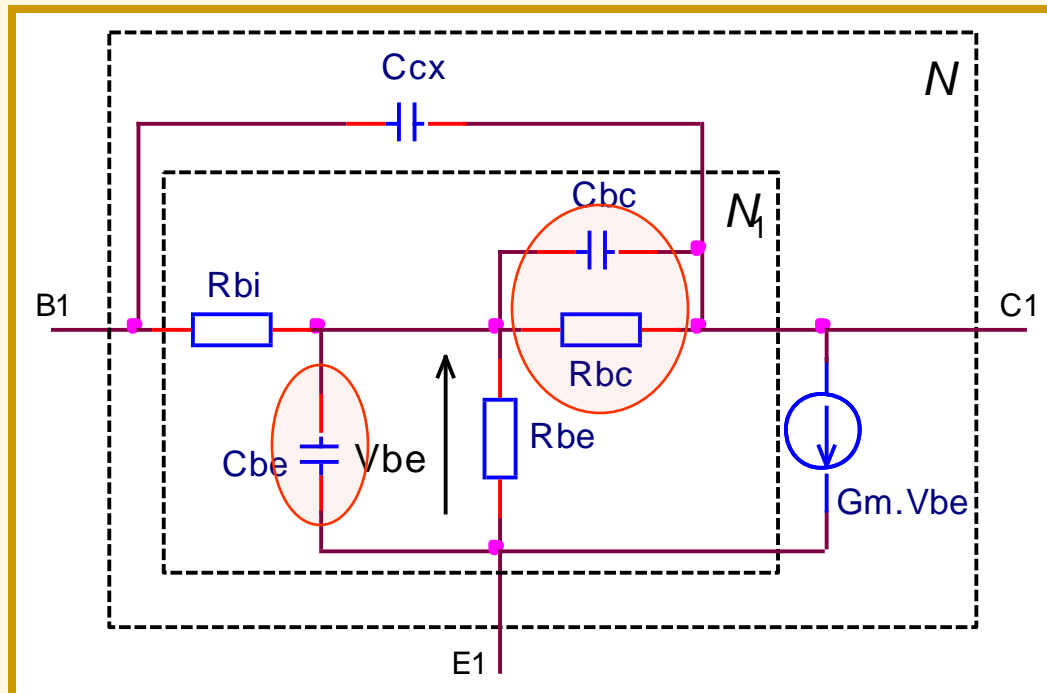
The following macros are defined for determination of  $R_{be}$  and  $R_{bi}$ :

$$C = 1 + R_{bi}C_{bc}$$

$$R_{be} = R_{bc} \cdot (T - T_{be}) / (R_{bi}C_{bc} - C \cdot T_{be})$$

$$R_{bi} = R_{bc} - C \cdot R_{be}$$

## ***F. Determination of $R_{bc}$ , $C_{bc}$ and $C_{be}$***



$$R_{bc} = \frac{R_{bi}}{R_{bibc}} \quad ; \quad C_{bc} = \frac{R_{bi} C_{bc}}{R_{bi}} \quad ; \quad C_{be} = \frac{T_{be}}{R_{be}} . \quad (24)$$

The following macros are defined for determination of Rbc, Cbc and Cbe:

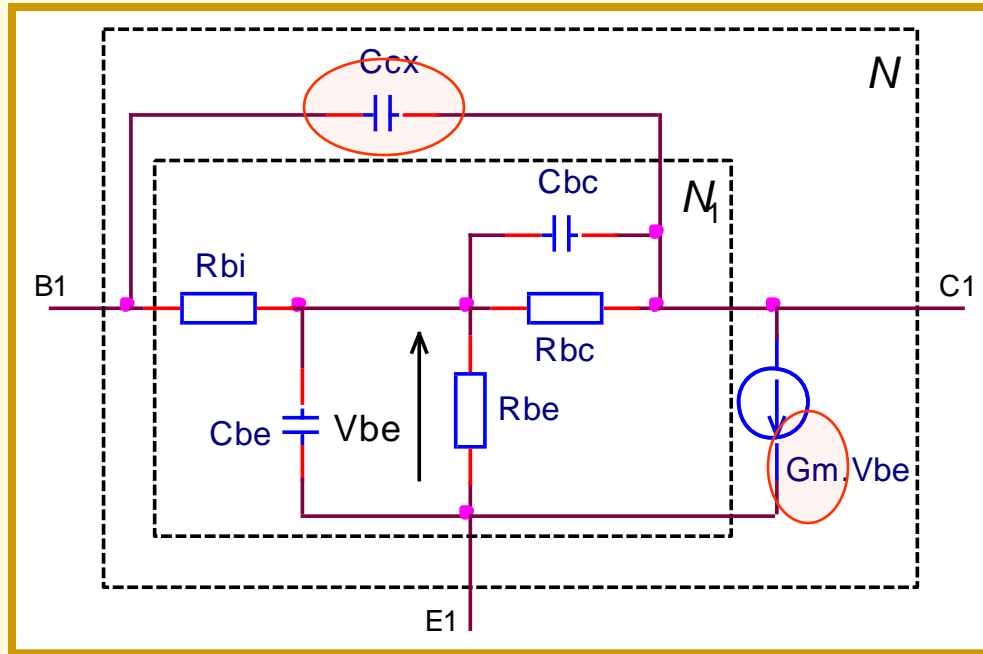
$$Rbc = Rbi/Rbibc$$

$$Cbc = RbiCbc/Rbi$$

$$Cbe = Tbe/Rbe$$



# Determination of $C_{cx}$ , $G_{mo}$ and $\tau$



$$Z_2 = \frac{Z_1 R_{bc}}{R_{be}}$$

$$Z_{cx} = \frac{Z_4 Z_2}{Z_2 - Z_4}$$

$$C_{cx} = \frac{1}{j\omega Z_{cx}}$$

$$Z_2 = Z_1 * R_{bc} / R_{be}$$

$$Z_3 = Z_1 / K Z_{13}$$

$$Z_4 = -1 / Y_{12}$$

$$Z_{cx} = Z_2 * Z_4 / (Z_2 - Z_4)$$

$$C_{cx} = \min(\text{img}(1 / (\omega * Z_{cx})))$$



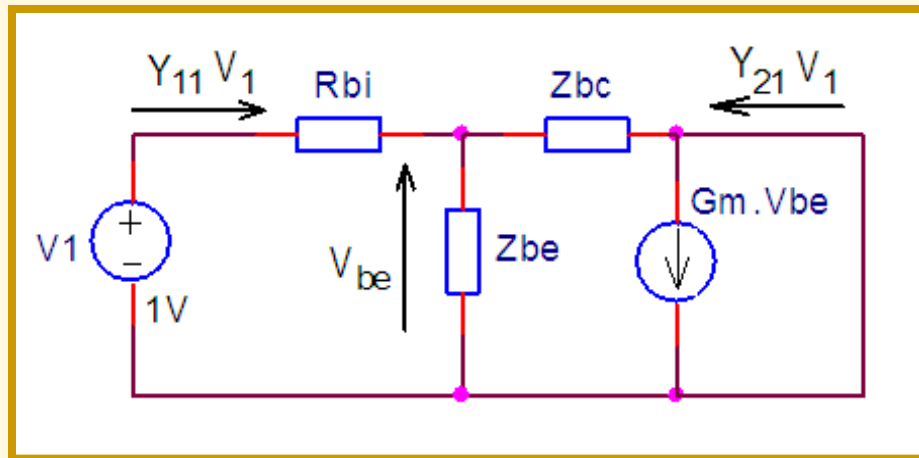


Fig. 5. Equivalent circuit for determination of  $G_m$

$$V_{R_{bi}} = y_{11} \cdot V_1 \cdot R_{bi}; \quad y_{21} \cdot V_1 = G_m \cdot V_{be} - \frac{V_{be}}{Z_{bc}}$$

$$V_{be} = V_1 - V_{R_{bi}} = V_1 (1 - y_{11} \cdot R_{bi});$$

$$G_m = G_{m0} e^{-j\omega\tau}.$$

$$G_m = \frac{y_{21}}{1 - y_{11} \cdot R_{bi}} + \frac{1}{Z_{bc}}$$

$$\tau = -\frac{\text{phase}(G_m)}{\omega}$$

## Macros in *Probe*

$$Gm\_f = Y_{21} / (1 - Y_{11} * R_{bi}) - 1 / Z_{bc}$$

$$Gm0 = \max(M(Y_{21} / (1 - Y_{11} * R_{bi}) - 1 / Z_{bc}))$$

$$TAU = \max(TAU\_f)$$

$$TAU\_f = (Frequency / F_{max}) * \text{rms}(-p(gm\_f) / w / (180 / \pi))$$





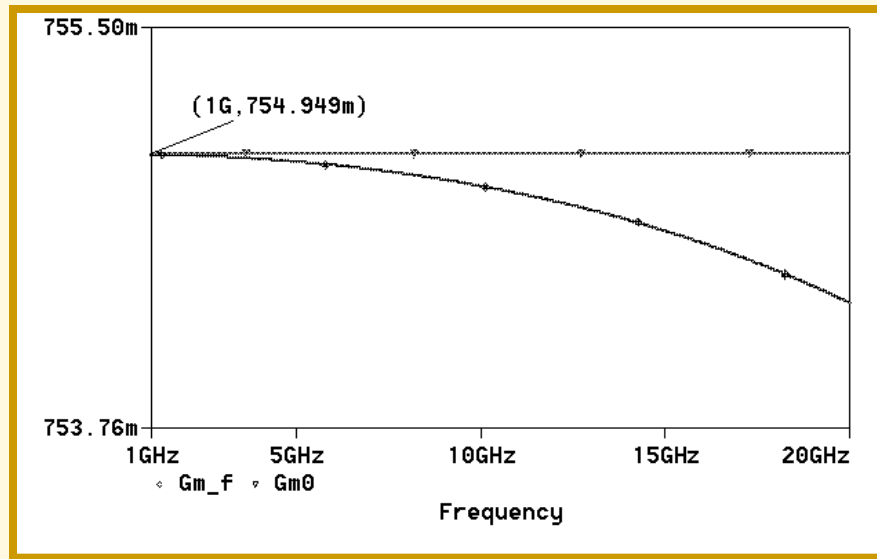


Fig. 6. Results for the extracted parameter  $Gm_0$

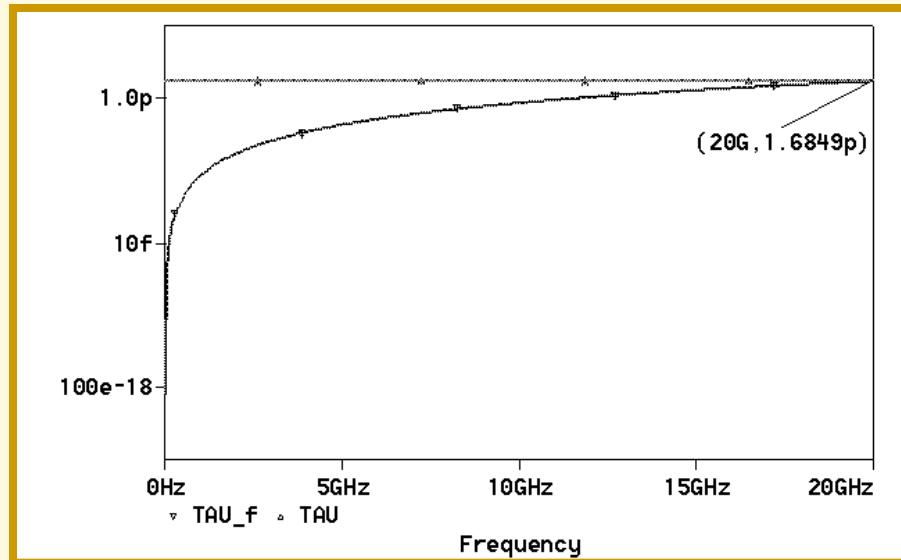


Fig. 7. Results for the extracted parameter  $\tau$

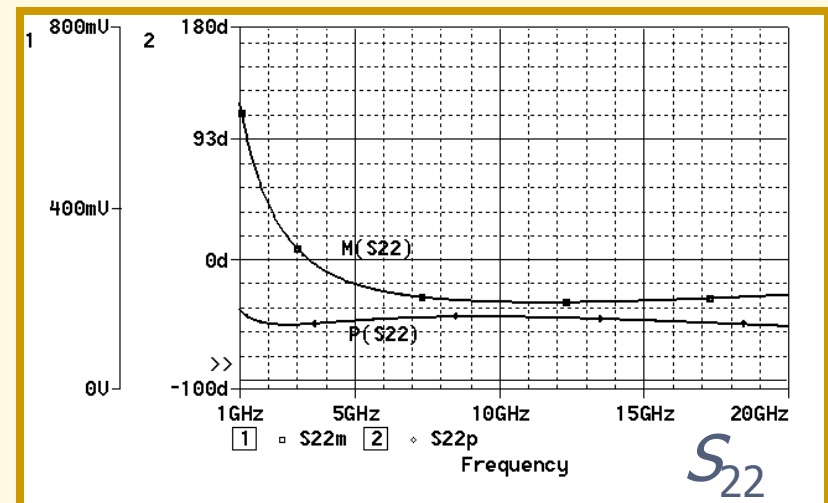
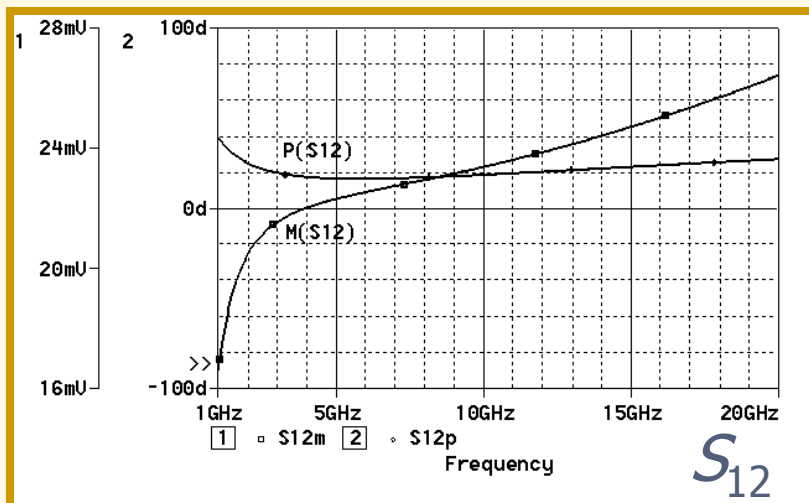
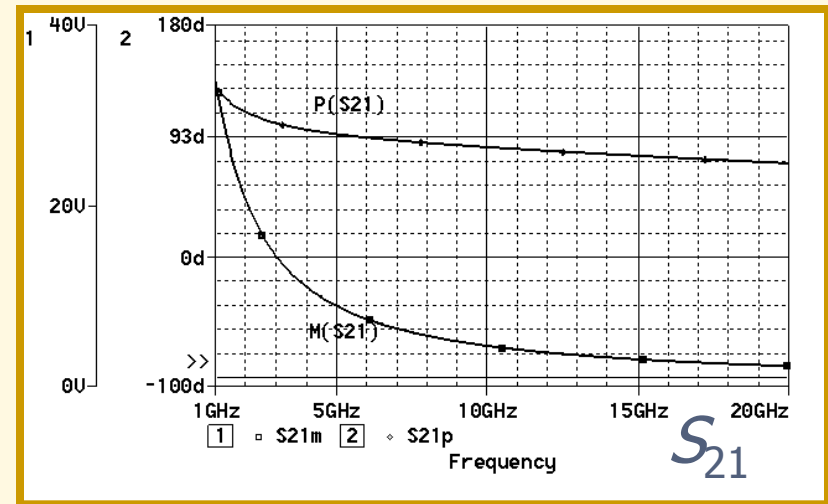
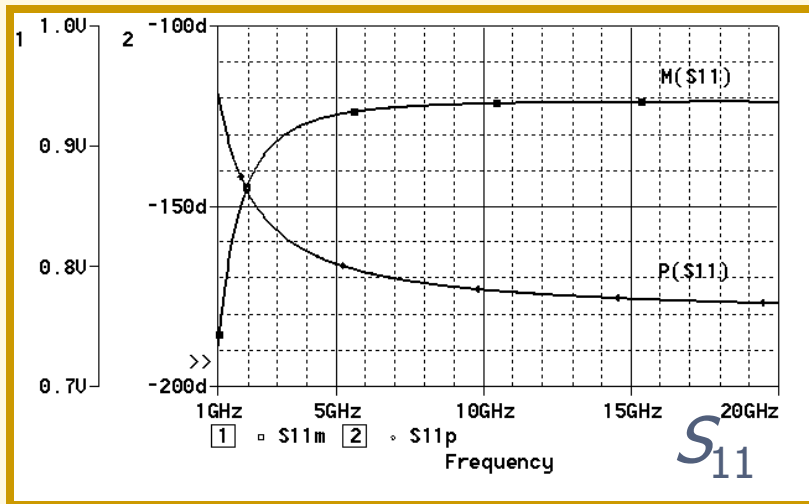


**Example:** Parameter extraction of small-signal HBT model based on measured data for  $S$ -parameters

TABLE II. EXTRACTED PARAMETERS OF THE SMALL-SIGNAL MODEL

Circuit elements	Extracted parameters	Extracted parameters in [1]
$R_{bi}$ [ $\Omega$ ]	1.9713	1.97
$R_{be}$ [ $\Omega$ ]	115.901	115.87
$R_{bc}$ [ $k\Omega$ ]	33.922	33.9
$C_{be}$ [fF]	2849.1	2849
$C_{bc}$ [fF]	6.3310	6.335
$C_{cx}$ [fF]	56.372	52.28
$G_{mo}$ [S]	0.754949	0.755
$\tau$ [ps]	1.6849	1.71

# Simulation results for the S-parameters



The results are in a good agreement with the measured S-parameters presented in [1] over the frequency range 1-20GHz. **The relative error does not exceed 5%.**

## CONCLUSIONS

- ❑ An approach to direct parameter extraction of small-signal HBT model has been developed, based on  $S$ -parameter measurements.
- ❑ The extraction procedure is realized in the Cadence Capture environment using the *Cadence PSpice* simulator and the graphical analyzer *Probe*. The computer realization is performed using macro-definitions in *Probe*.
- ❑ The obtained results are in a very good agreement with the measurement data.



**Thank you for your attention!**

